AMENDMENTS TO THE DRAWINGS

The attached sheets of drawings include changes to FIGS. 1, 2, and 3 labeling FIG. 1, FIG. 2, and FIG. 3 as --BACKROUND ART-- as described in the pages 1-6 of the Specification. Additionally, reference characters "$Lout_{(Inv)}$" and "$Lout_{(Non\text{-}inv)}$" have been added to distinguish the inverted and non-inverted Lout as shown in FIGS. 2, 3, 4, and 5, as described in page 5, line 21 – page 6, line 17 and page 11, line 24 – page 12, line 11. Also, "VSS~HVSS2" has been deleted from FIG. 4.

Attachment: Replacement sheet
Annotated sheet showing changes

REMARKS

This amendment is responsive to the Office Action dated October 20, 2008. Claim 1 is currently amended. New claim 3 has been added. Support for these amendments may be found variously throughout the Specification, for example on page 9, lines 26 – page 12, line 13. *These amendments add no new matter.* In the amendment, claim 1 remain pending in the application. Reconsideration and allowance of the pending claims are respectfully requested.

Claim 1 has been rejected under 35 U.S.C. § 112, ¶2, as being indefinite for failing to particularly point out and distinctly claim what Applicant regards as the invention.

Applicant appreciates the Examiner's attention to claim 1 in this regard, and has amended claim 1 to clarify essential structural cooperative relationships.

Applicant submits that the claim 1 is recited with the requisite particularity and distinctiveness, and respectfully requests reconsideration and withdrawal of the rejection of claim 1 as being indefinite under 35 U.S.C. § 112, ¶2.

Claim 1 has been rejected under 35 U.S.C. § 102(a) as being anticipated by what is referred to in the Office Action as Application's Admitted Prior Art (AAPA). This rejection is respectfully traversed.

Claim 1, as amended, recites: *[a] data transfer circuit for latching an input data in a first latch section, transferring a first latch result of said first latch section to a second latch section, and latching said first latch result in said second latch section, characterized by:*

transferring only an inverted output of said first latch result to said second latch section or transferring only a non-inverted output of said first latch result to said second latch section; and

raising a power supply voltage of said first latch section from a first voltage to a second voltage while said first latch result is transferred to said second latch section;

wherein said second voltage is higher than said first voltage and,

wherein said second voltage is a power supply voltage of said second latching section.

AAPA does not disclose or suggest these claimed features. It should be noted that, as disclosed by AAPA, "an inverted output $1Lout_{(Inv)}$ of a latch result by this first latch section

21 ***and*** a non-inverted output $1Lout_{(Non\text{-}inv)}$ thereof are inputted to the second latch section 22[.]" (Clean Substitute Spec., page 5, lines 21-28 (emphasis added).)

The Office Action infers that it would be possible for either the transfer switch 25 or the transfer switch 24 to "never" output an inverted or non-inverted output. (Office Action, pg. 13, lines 20-23.) However, AAPA clearly discloses that both transfer switches 25, 24 turn to an ON-state at a timing OE1. As shown in FIG. 2, timing OE1 turns transfer switches 24, 25 "ON-state". (Clean Substitute Spec., page 5, lines 25-27.) AAPA does not suggest that timing OE1 selectively turns one switch to an ON-state while the other transfer switch "never outputs" an output, either inverted or non-inverted.

Accordingly, AAPA fails to disclose or suggest "*transferring **only** an inverted output of said first latch result to said second latch section **or** transferring **only** a non-inverted output of said first latch result to said second latch section*[.]"

Additionally, while FIG. 3(C) and 3(D) discloses the inverted and non-inverted output voltages $1Lout_{(Inv)}$ and $1Lout_{(Non\text{-}inv)}$ rising and falling with timing SP, AAPA clearly fails to disclose or suggest "*raising a power **supply** voltage of said first latch section from a first voltage to a second voltage while said first latch result is transferred to said second latch section*[.]"

Because AAPA clearly fails to disclose or suggest each and every element set forth in claim 1, Applicant respectfully requests reconsideration and withdrawal of the rejection of claims 1 as being anticipated by Shishido under 35 U.S.C. § 102(a). *Verdegaal Bros. v. Union Oil Co. of California*, 814 F.2d 628, 631, 2 USPQ2d 1051, 1053 (Fed. Cir. 1987) ("A claim is anticipated only if each and every element as set forth in the claim is found, either expressly or inherently described, in a single prior art reference."); *Richardson v. Suzuki Motor Co.*, 868 F.2d 1226, 1236, 9 USPQ2d 1913, 1920 (Fed. Cir. 1989) ("The identical invention must be shown in as complete detail as is contained in the ... claim.")

Claim 1 has been rejected under 35 U.S.C. § 103(a) as being unpatentable over U.S. Pub. No. 2003/0011584 A1 to Azami et al. ("Azami"), or alternatively under 35 U.S.C. § 103(a) as being unpatentable over U.S. Pub. No. 2003/0011584 A1 to Azami et al. ("Azami") in view of Japanese Pat. No. 2000-221926 A to Nakajima et al. ("Nakajima"). This rejection is respectfully traversed.

Azami discloses first latch circuit and the second latch circuit used in the light emitting device wherein "a digital image signal is input (data in) from an input electrode of a TFT 1850, and a sampling pulse is input (pulse in) to a gate electrode, turning the TFT 1850 on, then the digital image signal is input to an inverter composed of TFTs 1851 to 1854 and a capacitor 1855, the polarity is inverted[,]" and "the digital image signal is also written into, and stored in, the second latch circuit by similar operations in accordance with a latch pulse (LAT) input timing." (Azami, para. [0183]-[0186].)

However, Azami clearly fails to disclose or suggest "*raising a power supply voltage of said first latch section from a first voltage to a second voltage while said first latch result is transferred to said second latch section;*

wherein said second voltage is higher than said first voltage and,

wherein said second voltage is a power supply voltage of said second latching section."

Nakajima fails to remedy the deficiencies of Azami. Nakajima discloses a CMOS latch cell 30 having CMOS inverters 31, 32. (Nakajima, para. [0037]-[0038], Drawing 5.) Nakajima discloses that "in the period of latch operation with active output enable pulse eo1, it operates under VDD1 power supply[.]" (Nakajima, para. [0043].) Further, Nakajima discloses that in the period of active output operation, "the power supply by the side of right of CMOW latch cell switches from VDD1 power supply to VDD2 power supply[,]" and that "power-supply-voltage VDD2 [is] higher than power-supply-voltage VDD1." (Nakajima, paras. [0044] and [0040].)

However, Nakajima fails to disclose "*raising a power supply voltage of said first latch section from a first voltage to a second voltage while said first latch result is transferred to said second latch section ...* ***wherein said second voltage is a power supply voltage of said second latching section***."

This feature is significant as Applicant's invention is intended to propose a simplified construction of a data transfer circuit. (Clean Substitute Spec., page 6, lines 18-27; page 15, line 31 – page 16, line 10.)

Accordingly, because Azami and Nakajima, either alone or in any permissible combination, fail to teach or suggest each and every feature of claim 1, Applicant respectfully

requests reconsideration and withdrawal of the rejection of claim 1 under 35 U.S.C. § 103(a) as being unpatentable over Azami, or in the alternative as being unpatentable over Azami in view of Nakajima. *In re Royka*, 490 F.2d 981, 180 USPQ 580 (CCPA 1974) (To establish prima facie obviousness of a claimed invention, all the claim limitations must be taught or suggested by the prior art.); *see also* MPEP 2143.03.

CONCLUSION

In view of the foregoing arguments, all claims are believed to be in condition for allowance. If any further issues remain, the Examiner is invited to telephone the undersigned to resolve them.

This response is believed to be a complete response to the Office Action. However, Applicant reserve the right to set forth further arguments supporting the patentability of their claims, including the separate patentability of the dependent claims not explicitly addressed herein, in future papers. Further, for any instances in which the Examiner took Official Notice in the Office Action, Applicant expressly do not acquiesce to the taking of Official Notice, and respectfully request that the Examiner provide an affidavit to support the Official Notice taken in the next Office Action, as required by 37 C.F.R. § 1.104(d)(2) and MPEP § 2144.03.

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Respectfully submitted,

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DESCRIPTION

DATA TRANSFER CIRCUIT AND FLAT DISPLAY DEVICE

BACKGROUND OF THE INVENTION

Technical field

The present invention relates to a data transfer circuit and a flat display device that can be applied to a liquid crystal display apparatus having a drive circuit formed integrally, for example, on an insulating substrate. According to the present invention, it is configured such that only an inverted output of a latch result of a first latch section or only a non-inverted output thereof is data-transferred to a second latch section, and that at least during a data transfer to the second latch section, a power supply voltage of the first latch section is raised, thereby enabling the simplification of a configuration relating to the data transfer, in a construction with TFTs and the like.

Background Art

Recently, in a liquid crystal display apparatus of a flat display device applicable to a portable terminal such as, for example, PDAs, portable telephones and the like, there has been provided an arrangement in which a drive circuit of a liquid crystal display panel is formed integrally on a glass substrate which is an insulating substrate constituting the liquid crystal panel.

Namely, FIG. 1 is a block diagram showing this type of a liquid crystal display apparatus. In this liquid crystal display apparatus 1, each pixel is formed with a

liquid crystal cell 2, a polysilicon TFT (Thin Film Transistor) 3 as a switching element of this liquid crystal cell 2 and a hold capacitance which is not shown, and a display section 4 of a rectangular shape is formed by arranging each pixel in a matrix. In this liquid crystal display apparatus 1, by disposing a color filter to each pixel formed in the display section 4 as described above, pixels R, G, B in red, green and blue colors are repeated sequentially and cyclically in a horizontal direction, with 240 sets of red, green and blue color pixels R, G and B as one set, and pixels in the horizontal direction are formed so as to form the display section 4. In this liquid crystal display unit 1, gradation data R0-R5, G0-G5, B0-B5, each with 6 bits, for indicating the gradation of these red, green, blue color pixels R, G, B are inputted simultaneously and in parallel in the order of raster scan, and the unit is enabled to display a desired image by driving each pixel on the basis of this gradation data D1 (R0-R5, G0-G5, B0-B5).

In the liquid crystal display apparatus 1, signal lines SL and gate lines SG of the display section 4 are connected to a horizontal drive circuit 5 and a vertical drive circuit 6, respectively. The horizontal drive circuit 5 outputs a drive signal for pixels corresponding to each signal line SL on the basis of the gradation data D1, and the vertical drive circuit 6, in correspondence to an output of the drive signal to the signal line SL by this horizontal drive circuit 5, selects pixels per line unit in the display section 4 by controlling the gate lines SG. Thereby, in the liquid crystal display

apparatus 1, a desired image is enabled so as to be displayed by driving each pixel in the display section 4 with these horizontal drive circuit 5 and vertical drive circuit 6.

More specifically, by selecting a plurality of reference voltages V0-V63 according to gradation data, the horizontal drive circuit 5, as described in Japanese Patent Application Publication No. 2000-242209, is enabled so as to perform a digital/analog conversion processing of the gradation data D1 and generate a drive signal. That is, in the horizontal drive circuit 5, by means of sampling latch circuits (SL) 8 that are provided corresponding to the disposition of pixels in the horizontal directions, by sequentially and cyclically sampling corresponding bits of R0-R5, G0-G5, B0-B5 of the gradation data D1, and by arranging together these gradation data D1 per line unit, they are outputted to reference voltage selectors 9 corresponding thereto. A reference voltage generating circuit 10 generates and outputs a plurality of reference voltages V0-V63 corresponding to each gradation of the gradation data D1. The reference voltage selector 9 selects the reference voltages V0-V63 outputted from this reference voltage generating circuit 10, based on the output data from each sampling latch circuit 8, and outputs a drive signal obtainable by a digital/analog conversion processing of corresponding gradation data D1. A buffer circuit 11 outputs this drive signal to a corresponding signal line SL.

FIG. 2 is a connection diagram showing a configuration for one bit portion of the sampling latch

circuit 8 in the horizontal drive circuit 5 composed as described above. In the sampling latch circuit 8, after latch-holding gradation data D1 in a first latch section 21 at a timing corresponding to a position of a corresponding pixel in the horizontal direction, a latch result of the first latch section 21 is transferred and outputted to a second latch section 22 at a predetermined timing set in a vertical blanking period, thereby arranging the gradation data together per line unit and outputting it to the reference voltage selector 9. Here, with respect to active elements, such as a low temperature polysilicon TFT or the like which is formed on an insulating substrate for constructing this type of sampling latch circuit 8 or the like, there is a large dispersion in their characteristics. Therefore, in the sampling latch circuit 8, the circuit is configured to output the latch result to the second latch section 22 with a so-called bi-phase output by outputting an inverted output $\underline{1Lout_{(Inv)}}$ of the latch result as well as a non-inverted output $\underline{1Lout_{(Non\text{-}inv)}}$ thereof so as to ensure a stable and reliable data transfer of the latch result to be secured between the first latch section 21 and the second latch section 22.

That is, in the first latch section 21 in this sampling latch circuit 8, a CMOS inverter composed of a N-channel MOS (hereinafter referred to as a NMOS) and a P-channel MOS (hereinafter referred to as a PMOS), a gate and a drain being connected respectively in common therebetween, as well as a CMOS inverter composed of an NMOS transistor Q3 and a PMOS transistor Q4, and likewise, a gate and a drain being connected respectively in common

therebetween, are provided in parallel between a positive side power supply line of power supply voltage VCC and a negative side power supply line of power supply voltage VSS. In the first latch section 21, an inverter output by means of the transistors Q1 and Q2 is inputted to an inverter composed of the transistors Q3 and Q4. Further, via a PMOS transistor Q5 operating at an inverted signal ~~xsp~~ xSP of a sampling pulse ~~sp~~ SP, an inverter output by means of the transistors Q3 and Q4 is inputted to the inverter composed of the transistors Q1 and Q2, and still further, via a PMOS transistor Q6 operating at a sampling pulse ~~sp~~ SP, gradation data D1 is inputted to the inverter composed of the transistors Q1 and Q2.

Thereby, in the sampling latch circuit 8, a CMOS latch cell with a comparator configuration is formed with transistors Q1 to Q6, wherein, as shown in FIGS. 3(A) to 3(D), gradation data D1 is caused to be latched by the sampling pulse ~~sp~~ SP, and a timing of this latch is configured to be set according to the position of a corresponding pixel in the horizontal direction.

In the sampling latch circuit 8, an inverted output $1Lout_{(Inv)}$ of a latch result by this first latch section 21 and a non-inverted output $1Lout_{(Non\text{-}inv)}$ thereof are inputted to the second latch section 22, respectively, via transfer switches ~~24, 25~~ 25, 24. Here, this transfer switch ~~24, 25~~ 25, 24 turns to an ON-state at a timing OE1, for example, of a rise timing in a horizontal blanking period ~~(FIG. 9 (E))~~ (FIG. 3(E)).

In the second latch section 22, a latch cell is formed with a CMOS inverter composed of an NMOS transistor Q7 and a PMOS transistor Q8, and a CMOS

inverter composed of an NMOS transistor Q9 and a PMOS transistor Q10. An inverted output $1Lout_{(Inv)}$ and a non-inverted output $1Lout_{(Non\text{-}inv)}$ of the latch result inputted via the transfer switches ~~24, 25~~ 25, 24 are inputted to the CMOS inverter composed of the transistors Q7, Q8 and the CMOS inverter composed of the transistors Q9, Q10, respectively. Thereby, the sampling latch circuit 8 is configured to perform data transfer of a latch result of the first latch section 21 at a timing OE1 of the rise in the horizontal blanking period for latching in the second latch section 22 (FIG. 3.(F)), and to output ~~this~~ the output 2Lout of the latch result of the second latch via an inverter 26. By way of example, in the second latch section 22, by setting a positive power supply and a negative power supply appropriately, a latch output may be outputted after level-shifting to be suitable for processing in the following reference voltage selector 9.

However, in the case of data transfer of the latch result and the like by use of the bi-phase as described above, there is a problem that its configuration becomes complicated in comparison with a data transfer by use of a single phase. If the construction relating to such data transfer can be simplified, an overall configuration can be simplified in accordance therewith, and a so-called narrow bezel can be realized in this type of display apparatus. Further, power consumption can be reduced.

DISCLOSURE OF THE INVENTION

The present invention has been contemplated in consideration of the problems described above, and the

invention is intended to propose a data transfer circuit and a flat display apparatus in which the construction relating to data transfer in the configuration thereof with TFTs and the like can be simplified.

In order to solve the problems described above, the present invention is applied to a data transfer circuit in which an input data is latched in a first latch section, and a latch result of the first latch section is data-transferred to a second latch section for latching therein, in which only an inverted output of the latch result of the first latch section or only a non-inverted output of the latch result is allowed to be data-transferred to the second latch section, and in which at least during the period of data transfer of the latch result from the first latch section to the second latch section, a power supply voltage of the first latch section is caused to rise.

According to the configuration of the present invention, if only the inverted output of the latch result of the first latch section or only the non-inverted output of the latch result is data-transferred to the second latch section, the configuration thereof can be simplified accordingly in comparison with the case of the data transfer of the latch result using both of the inverted output and the non-inverted result. Further, if the power supply voltage of the first latch section is enabled to be raised at least during the period of the data transfer of the latch result from the first latch section to the second latch section, it becomes possible to expand a margin in the data transfer, thereby enabling a compensation for a decrease in the margin due to the

data transfer of the latch result to the second latch section only by use of the inverted output or only by use of the non-inverted output of the latch result, by the merit of the expanded margin, and thereby ensuring a stable and reliable data transfer of the latch result.

Still further, the present invention is applied to a flat display apparatus, which includes: a plurality of latch circuits for sampling gradation data sequentially and cyclically, and distributing the gradation data to a corresponding column; and a digital/analog conversion circuit for setting an output signal level to the corresponding column depending on the latch result of the latch circuit, in which each latch circuit is configured to data-transfer only an inverted output of a latch result of a first latch section or only a non-inverted output of the latch result of the first latch section to a second latch section, and at least during the period of data transfer of the latch result of the first latch section to the second latch section, a power supply voltage of the first latch section is raised.

Thereby, according to the configuration of the present invention, in the latch circuit of the flat display apparatus, the circuit is enabled to perform data transfer of the latch result stably and reliably with a simplified configuration.

BRIEF DESCRIPTION OF DRAWINGS

FIG. 1 is a block diagram showing a configuration of a liquid crystal display apparatus.

FIG. 2 is a connection diagram showing a sampling latch circuit applied to a conventional liquid crystal

display apparatus.

FIG. 3 is a time chart for use in the description of the operation of the sampling latch circuit shown in FIG. 2.

FIG. 4 is a connection diagram showing a sampling latch circuit according to an embodiment of the present invention.

FIG. 5 is a time chart for use in the description of the operation of the sampling latch circuit shown in FIG. 4.

BEST MODE FOR CARRYING OUT THE INVENTION

A preferred embodiment of the present invention will be described in detail in the following by referring to the accompanying drawings.

(1) First Embodiment

FIG. 4 is a connection diagram showing, in comparison with that shown in FIG. 2, a configuration for one bit portion of a sampling latch circuit to be applied to a liquid crystal display apparatus according to an embodiment of the present invention. Because this liquid crystal display apparatus according to the embodiment of the present invention is configured the same as the liquid crystal display apparatus 1 described with reference to FIGS. 1, 2, except that a configuration of this sampling latch circuit 38 differs therefrom, a duplicative description will be omitted.

In this sampling latch circuit 38, after latching gradation data D1 by the first latch section 41 at a timing corresponding to the disposition of pixels in the horizontal direction, a latch result by this first latch

section 41 is transferred to the second latch section 42 at a predetermined timing OE1 in a horizontal blanking period for latching therein, and outputting to a reference voltage selector 9 subsequent thereto. This sampling latch circuit 38 executes data transfer of the latch result from the first latch section 41 to the second latch section 42 by use of a single phase, and secures a margin which becomes deficient due to the single phase data transfer to be compensated by raising the power voltage.

That is, in this sampling latch circuit 38, the first latch section 41 is provided with a CMOS inverter composed of an NMOS transistor Q11 and a PMOS transistor Q12 as well as a CMOS inverter composed of an NMOS transistor Q13 and a PMOS transistor Q14 arranged in parallel between a positive side power supply VH and a negative power supply VSS. In the first latch section 41, an inverter output from the transistors Q11 and Q12 is inputted to the inverter composed of the transistors Q13 and Q14; further, via a switching circuit 44 which operates off-action at a sampling pulse ~~sp~~ SP, an inverter input from the transistors Q11 and Q12 is inputted to the inverter composed of the transistors Q13 and Q14, and still further, via a PMOS transistor Q15 which operates on-action at a sampling pulse ~~sp~~ SP, gradation data D1 is inputted to the inverter composed of the transistors Q11 and Q12.

Thereby, in the sampling latch circuit 38, a CMOS latch cell is formed with the transistors Q11 to Q15, and as shown in FIGS. 5 (A)-(C), it is configured such that, after fetching gradation data D1 while setting the switch

circuit 44 off-state by a sampling pulse ~~sp~~ SP, the switch circuit 44 is set to an ~~set~~ on-state to hold the gradation data D1 fetched in, and a timing relating to this latch is set in accordance with the position of a corresponding pixel in the horizontal direction.

Furthermore, in the first latch section 41, by selection of a power supply via a switch circuit 47, a processing relating to this latch is executed in a state of setting to a power supply VDD1 of 2.9 V, which is equal to a power supply relating to a pre-stage circuit. Further, immediately before the data transfer of a latch result to the second latch section 42, a power supply VDD2 at 5.8V, which is higher than the voltage at the time of the latching, is selected, and then upon completion of the data transfer, the original power supply VDD1 is selected. Thereby, in this sampling latch circuit 38, at least during the period of data transfer of a latch result from the first latch section 41 to the second latch section 42, the circuit is configured to raise the power supply voltage to secure a margin which drops due to the data transfer of the latch result by use of the single phase.

Thereby, the first latch section 41 is configured, at a predetermined timing OE1 in a horizontal blanking period, so as to transfer the latch result by expanding the amplitude thereof to the second latch section 42 via a transfer switch 45. In this embodiment, an inverted output $1Lout_{(Inv)}$ is applied to the latch result to be supplied to this data transfer (FIGS. 5 (C) to (E)).

The second latch section 42 is provided with a CMOS inverter composed of an NMOS transistor Q16 and a PMOS

transistor Q17 as well as a CMOS inverter composed of an NMOS transistor Q18 and a PMOS transistor Q19 arranged in parallel between a positive side power supply VDD2 and a negative side power supply VL, thereby forming a latch cell in a comparator circuit construction by these CMOS inverters, and then an output from the transfer switch 45 is supplied to this latch cell. Thus, the second latch section 42 is configured to latch the latch result from the first latch section 41, and then to output the output 2Lout of ~~this~~ the latch result of the second latch via an inverter 46.

Further, the second latch section 42 is enabled so as to level shift a latch output to be outputted so that the output becomes suitable for processing in the reference voltage selector 9, by setting-up of the negative side power supply VL.

(2) Operation of the Embodiment

According to the configuration described above, in the liquid crystal display apparatus (FIG. 1), gradation data D1 composed of a series of data indicating the gradation of each pixel for use of the display are inputted to the horizontal drive circuit 5, in which these gradation data D1 are sampled sequentially by the sampling latch circuit 38 and arranged per line unit, and then by the reference voltage selector 9 subsequent thereto, a reference voltage V0-V63 is selected in accordance with each gradation data. In the liquid crystal display apparatus 1, a drive signal for driving each pixel is generated by selection of this reference voltage V0 to V63, and this drive signal is supplied to the display unit 4 via the signal line SL, and the drive

signal is applied to a pixel that is selected by the vertical drive circuit 6. Thereby, in the liquid crystal display apparatus 1, a desired image is displayed by driving each pixel in the display unit 4 in accordance with corresponding gradation data D1.

In the horizontal drive circuit 5 which drives the display section 4 as described above, in the sampling latch circuit 38 which sequentially and cyclically samples the gradation data D1 as described above (FIG. 4), after each bit of the gradation data D1 being latched in the first latch section 41 at a timing corresponding thereto, the bit is transferred simultaneously and in parallel to the second latch section 42 so as to be latched therein at a predetermined timing OE1 in a horizontal blanking period via each sampling latch circuit 38, and ~~this~~ the output 2Lout of the latch result of the second latch is outputted to the reference voltage selector 9. Thereby, in the liquid crystal display apparatus 1, the gradation data D1 is arranged per line unit, and subjected to a digital/analog conversion processing by the reference voltage selector 9.

In the sampling latch circuit 38, the data transfer from the first latch section 41 to the second latch section 42 is carried out by use of the inverted output $1Lout_{(Inv)}$ of the latch result as described above, thus performing the data transfer of the latch result by use of a single phase, and thereby, in comparison with the case of the data transfer by use of a bi-phase, enabling the simplification of the configuration. Specifically, in the transfer switch relating to such data transfer, a minimum of two transistors are necessary in the

construction of an inverter. In contrast, in the case of data transfer by use of a single phase as described above, especially in this embodiment, transfer switches for as many portions as 240 pairs × 3 (red, green, blue) × 6 bits can be omitted, thereby enabling the omission of as many pieces of transistors as 4320 × 2 in comparison with the case of the data transfer by use of the bi-phase. Thereby, according to this liquid crystal display apparatus, it becomes possible to decrease power consumption through simplification of the configuration and, further, to realize the so-called narrow bezel.

Further, during the period of such data transfer, in the first latch section 41, its power voltage is caused to rise, thereby securing a margin that drops during the data transfer of the latch result by use of the single phase. Thereby, in the liquid crystal display apparatus, the data of the latch result is configured so as to be transferred by use of a single phase, ~~tus~~ thus enabling the latch result to be transferred to the second latch section 42 stably and reliably.

(3) Advantage of the Embodiment

According to the configuration described above, by an arrangement such that only the inverted output $1Lout_{(Inv)}$ of the latch result in the first latch section is data-transferred to the second latch section and, the at least during the period of the data transfer to the second latch section, the power supply voltage of the first latch section is raised, the construction relating to the data transfer, in its configuration using TFTs, can be simplified.

(4) Other Embodiments

By way of example, in the description of the above embodiment, the case of the data transfer only of the inverted output $\underline{1Lout_{(Inv)}}$ of the latch result from the first latch section to the second latch section has been described, however, the present invention is not limited thereto, and it can be applied widely to a case of data transfer only of a non-inverted output of the latch result to the second latch section.

Further, in the description of the above embodiment, the case where the present invention is applied to the TFT liquid crystal which comprises the display section and the like formed on a glass substrate has been described, however, the present invention is not limited thereto, and it can be applied widely to various types of liquid crystal display apparatuses, such as CGS (Continuous Grain Silicon) liquid crystal and the like, and further to various flat display apparatuses, such as an EL (Electro Luminescence) display apparatus and the like.

Still further, in the description of the above embodiment, the case where the present invention is applied to the liquid crystal display apparatus, in which the first and the second latch sections are constructed with active elements composed of low temperature poly-silicon TFTs formed on the insulation substrate was described; however, the present invention is not limited thereto, and it can be applied widely to any data transfer circuit for carrying out data transfer in which the first and the second latch sections are formed with various active elements formed on an insulation substrate.

As described hereinabove, according to the present

invention, by configuring the invention such that the data of only the inverted output of the latch result of the first latch section or only the non-inverted output thereof is enabled to be transferred to the second latch section and, at least during the period of the data transfer to the second latch section, the power supply voltage of the first latch section is raised, in the configuration with the TFTs and the like, to simplification of the construction relating to the data transfer is enabled.

~~INDUSTRIAL APPLICABILITY~~

The present invention relates to a data transfer circuit and a flat display apparatus, and it can be applied to a liquid crystal display apparatus in which a drive circuit is formed integral, for example, on an insulation substrate.